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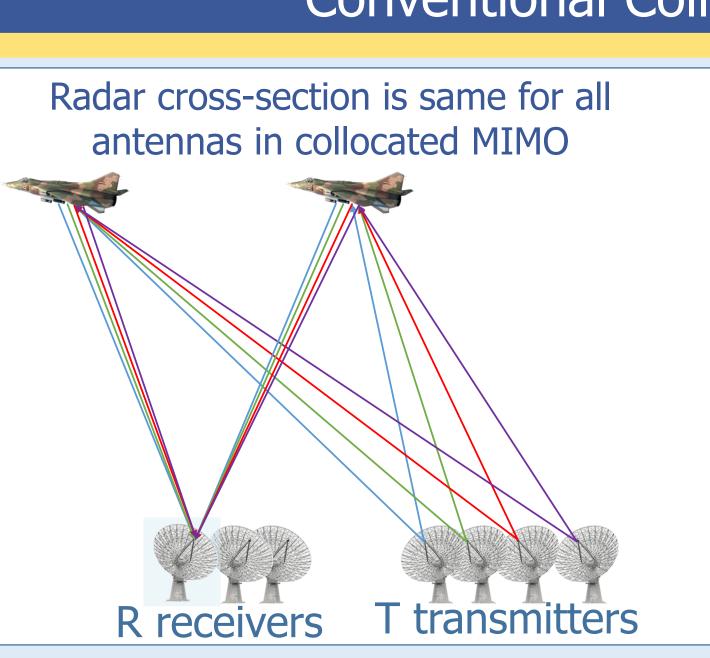




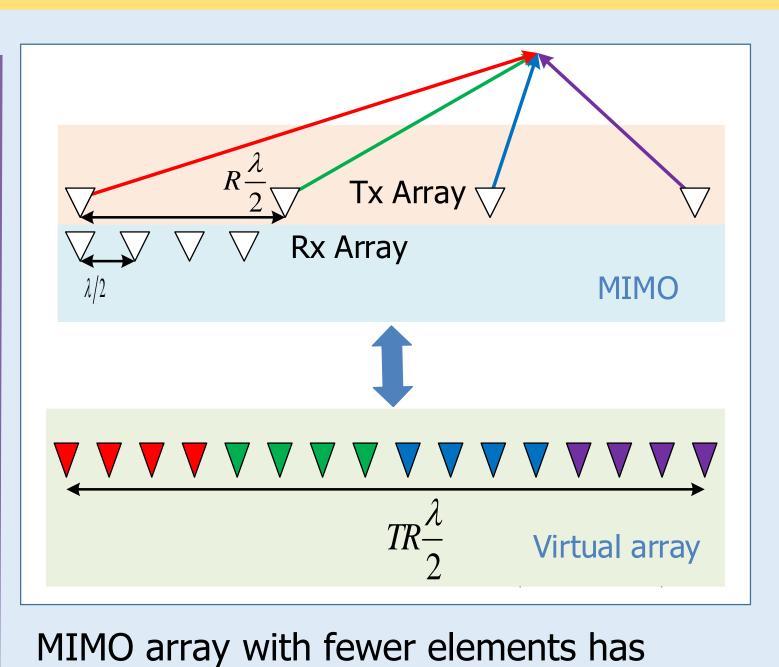


Theoretical Background

Conventional Collocated MIMO Radar



MIMO transmits orthogonal waveforms and processes linear combination of echoes received due to each waveform

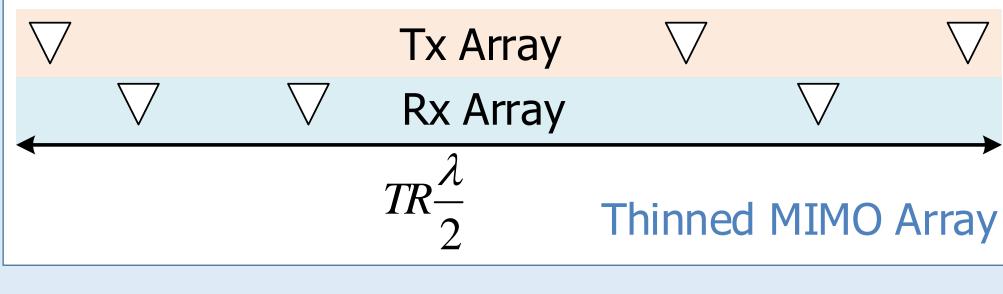


same spatial resolution as a virtual array with more elements

Sub-Nyquist Collocated MIMO Radar

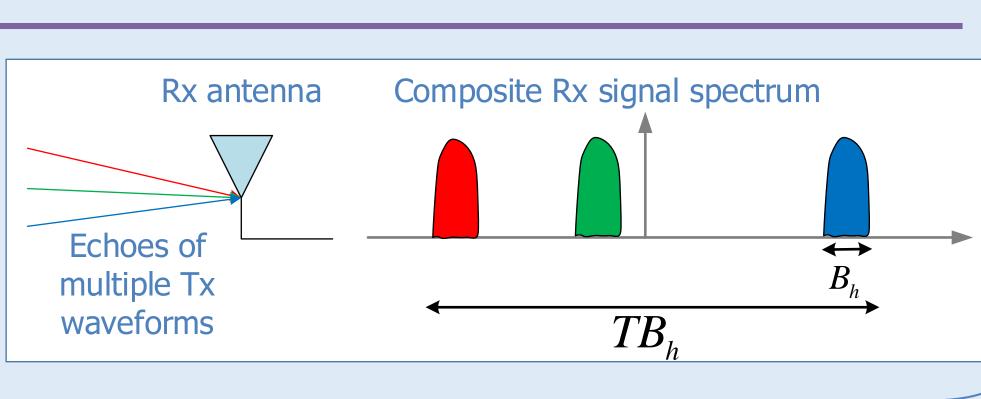
Spatial Sub-Nyquist

Thinned random array that preserves the azimuthal resolution as a virtual ULA



Spectral Sub-Nyquist

Reduced sampling rate at each receiver that preserves the range resolution as with bandwidth TB_h

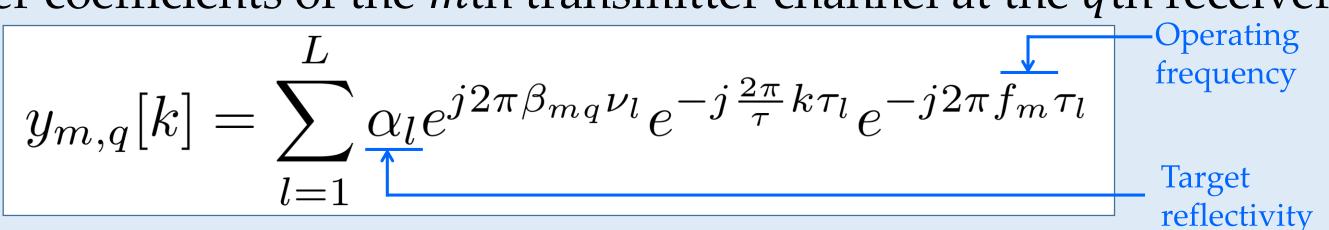


Signal Model and Xampling

Received signal at the qth antenna after demodulation:

$$x_q(t) = \sum_{m=0}^{M-1} \sum_{l=1}^{L} h_m(t-\tau_l) e^{j2\pi\beta_{mq}\nu_l} \text{ Total number of targets}$$

Element location factor Target time delay Fourier coefficients of the *m*th transmitter channel at the *q*th receiver:



Xampling retrieves the Fourier coefficients from low rate samples

Recovery algorithm: Matrix OMP

- Goal: Recover delay, azimuth and reflectivity from $y_{m,q}[k]$
- In matrix form, the Fourier coefficients for the *m*th transmission:

$$oldsymbol{Y}^m = oldsymbol{A}^m oldsymbol{X} (oldsymbol{B}^m)^T$$

Solve Range dictionary Azimuth dictionary

minimize
$$\| \boldsymbol{X} \|_0$$
 Sparse reflectivity matrix; non-zero values at target location

Use OMP for simultaneous sparse matrix recovery

Clutter Model

The received signal $r_q(t)$ at the q^{th} antenna is $r_q(t) = x_q(t) + y_q(t) + n_q(t)$ where $n_q(t)$ is the spatially and temporally white Gaussian noise.

The echo from *C* clutter targets is

 $y_{q}(t) = \sum_{n=0}^{\infty} \sum_{m=0}^{\infty} \sum_{c=1}^{\infty} \alpha_{c} h_{m}(t - p\tau - \tau_{c}) e^{j2\pi\beta_{m,q}\phi_{c}} e^{j\nu_{v}p\tau}$

• The clutter amplitude in the angular range $[\theta - \epsilon, \theta + \epsilon)$ is

 The clutter correlation function is

 $E[S_c(\theta)S_c^*(\theta')]=R_{S_c}(\theta-\theta')$

After normalization becomes

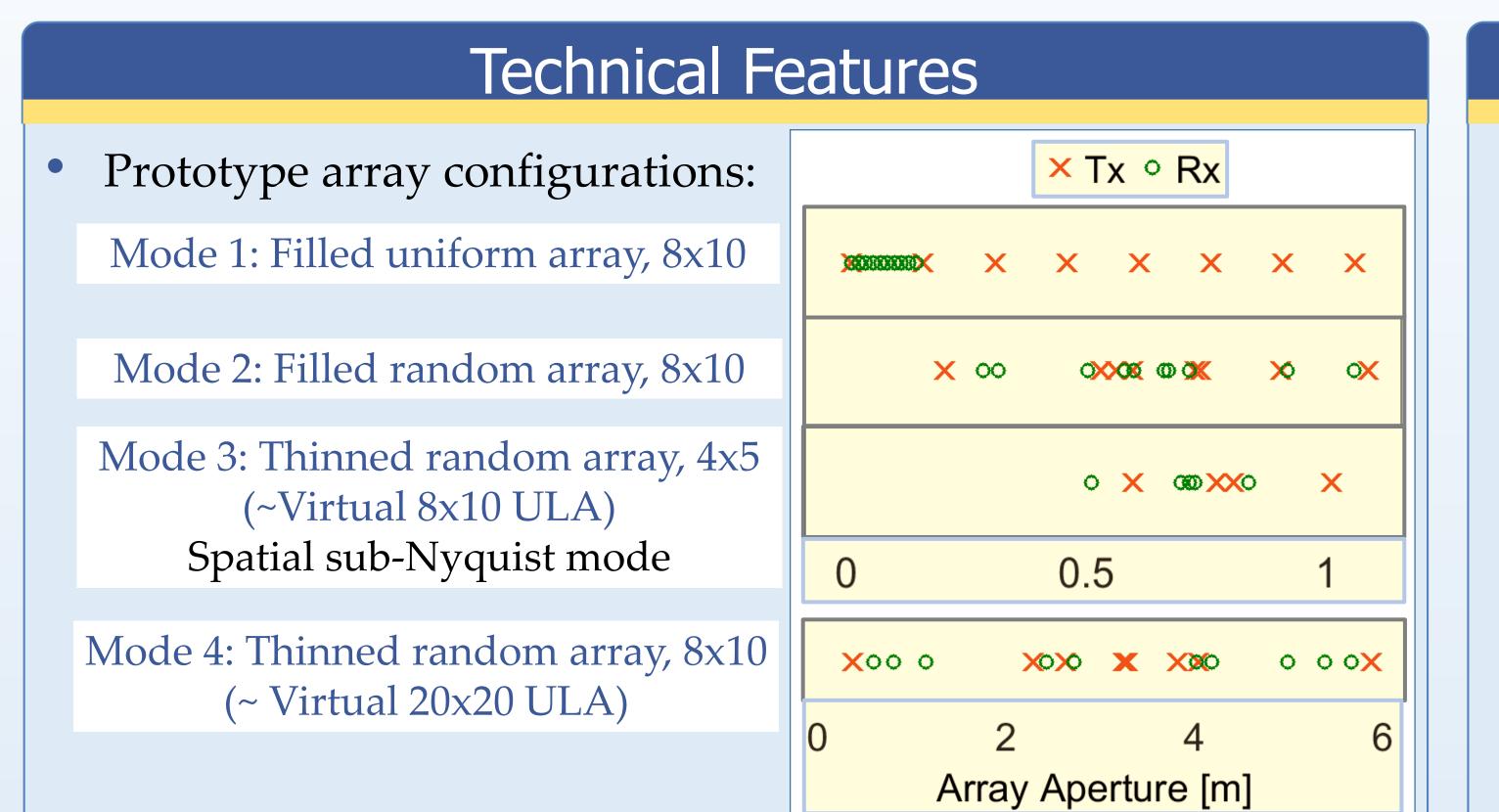
 $Y_{m,q,p}[k] = \frac{\tau}{\left| H_m \left(\frac{2\pi}{\tau} k \right) \right|^2} \tilde{Y}_{m,q,p}[k + f_m \tau] = \sum_{c=1} \alpha_c e^{j2\pi\beta_{m,q}\phi_c} e^{-j\frac{2\pi}{\tau}k\tau_c} e^{-j2\pi f_m \tau_c} e^{j\nu_c p\tau}$



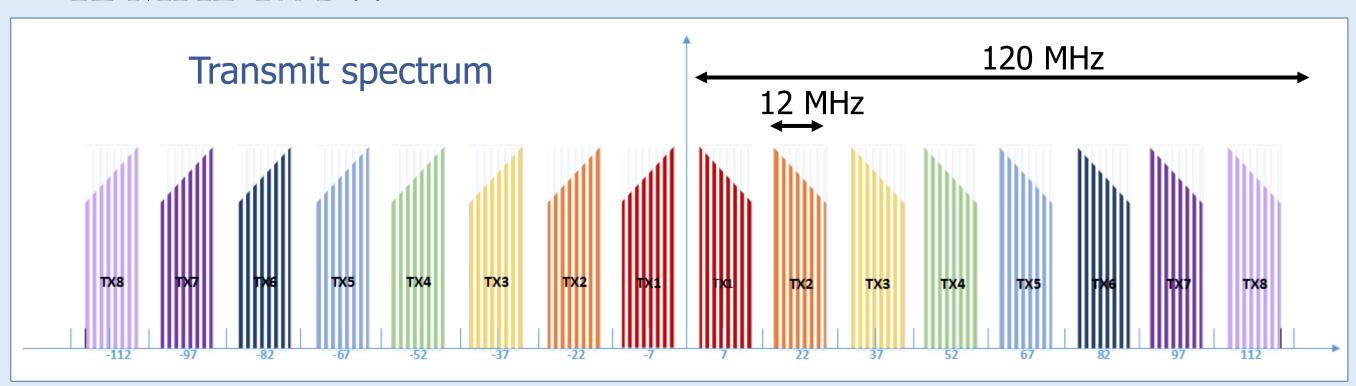




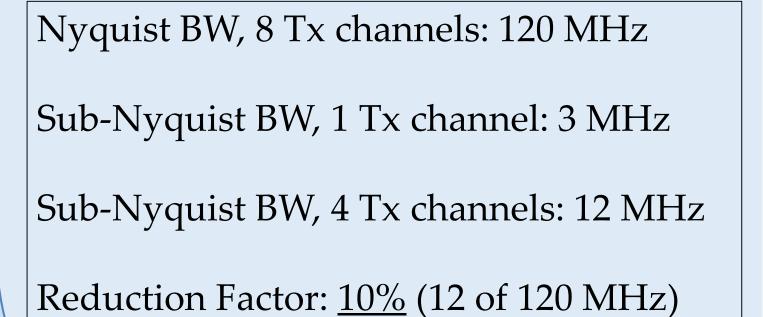
Cognitive Sub-Nyquist Collocated MIMO Radar Prototype with Clutter Removal

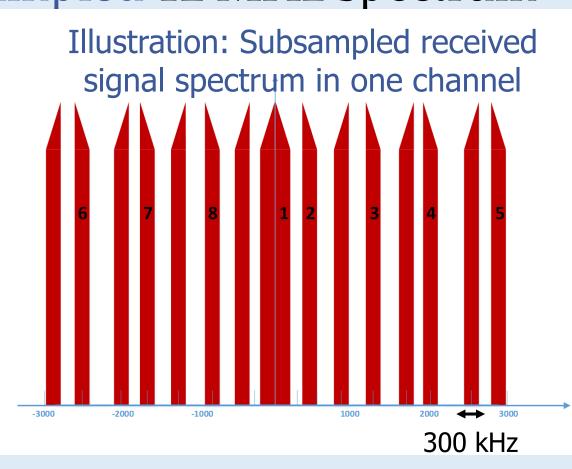


Cognitive transmission of eight 300 kHz bands within each 12 MHz Tx BW



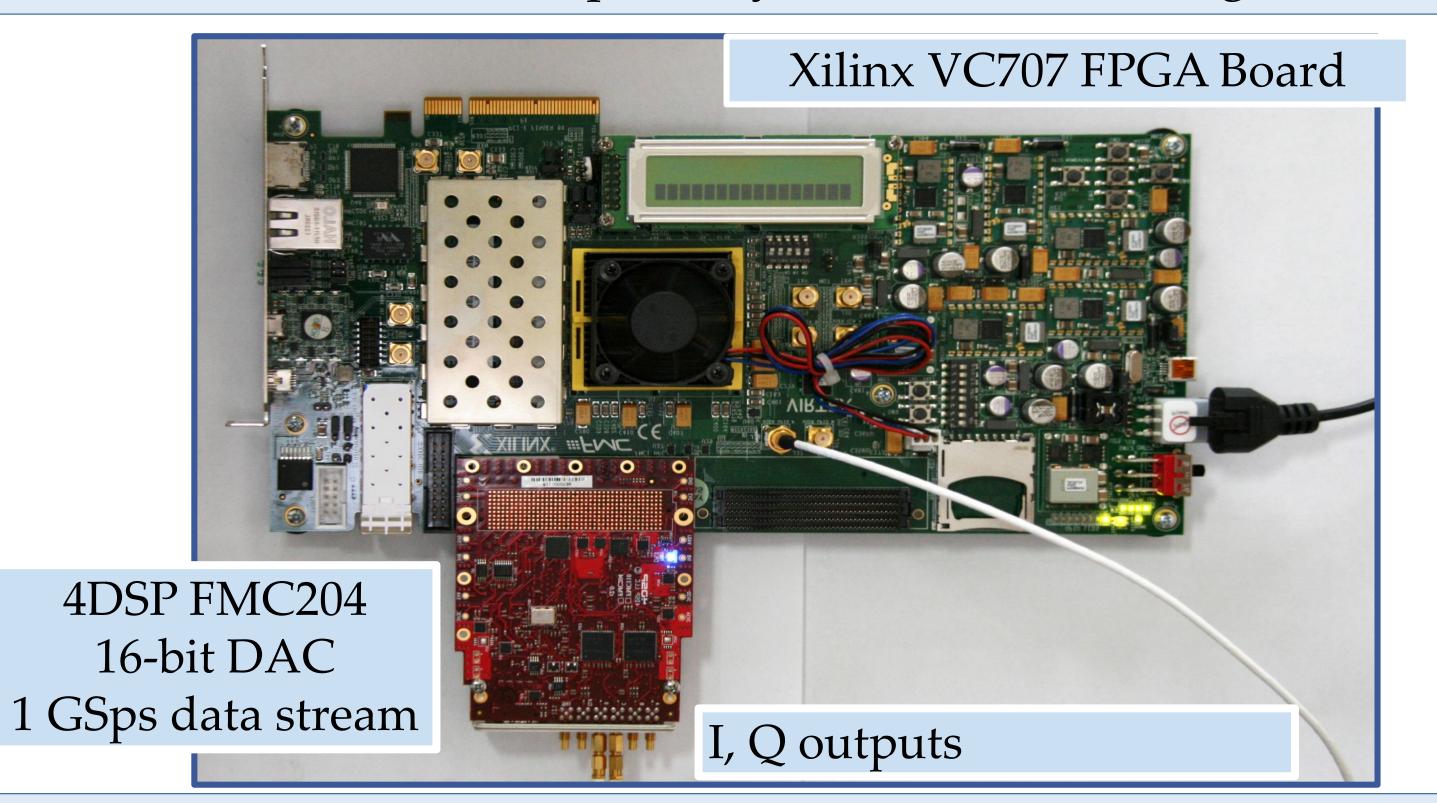
Sub-Nyquist processing of subsampled 12 MHz spectrum



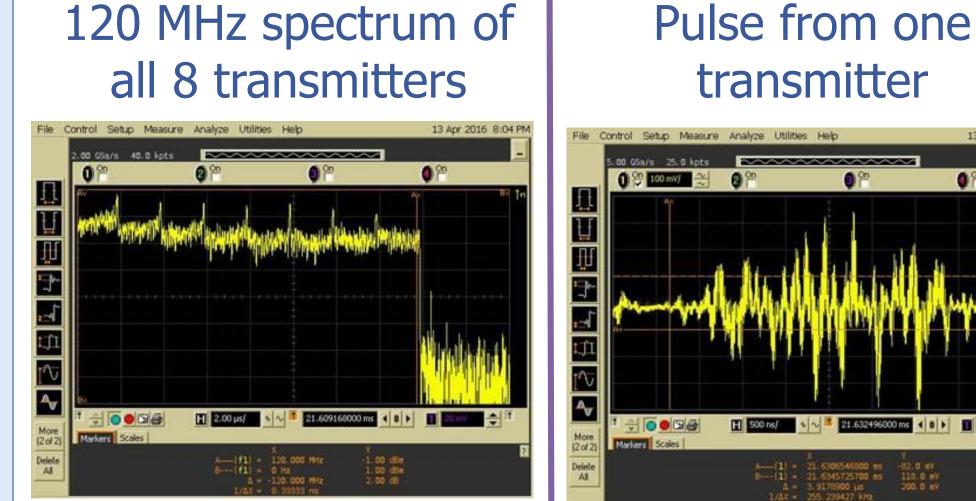


Transmit Waveform Generator

Virtex-7 XCVX486T FPGA based digital waveform generator serializes all receivers separately into I and Q analog channels



Transmitter outputs



Transmit signals differ in carrier frequencies

transmitter

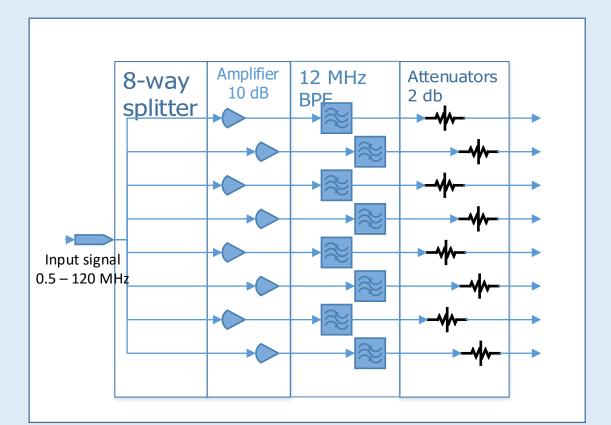
Spectrum of one transmitter

Only eight 300 kHz slices are transmitted

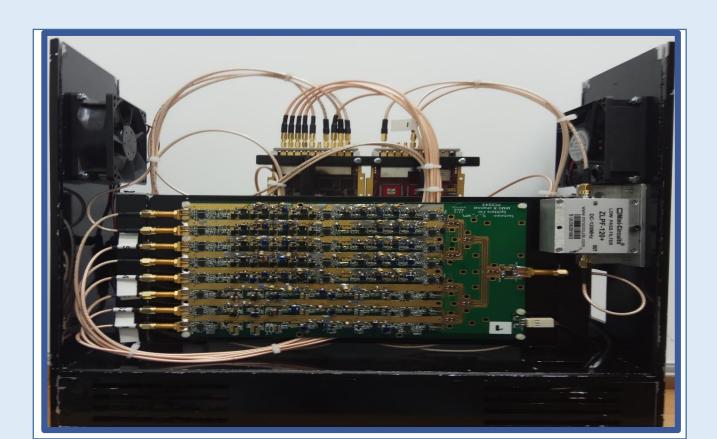
3 MHz guardband between adjacent channels

Analog Pre-Processor (APP) Board

APP filters the receiver data into eight channels



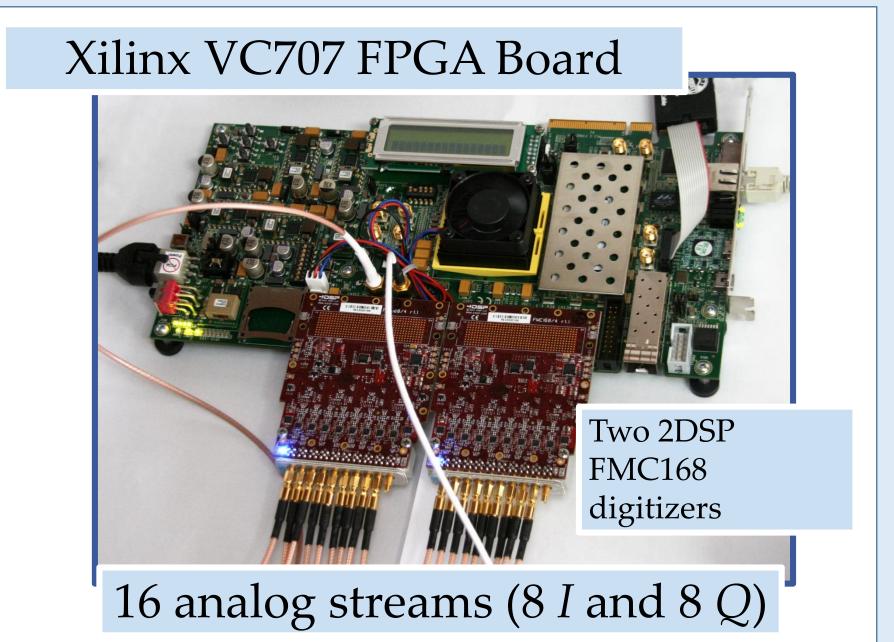
The APP card is mounted in a chassis



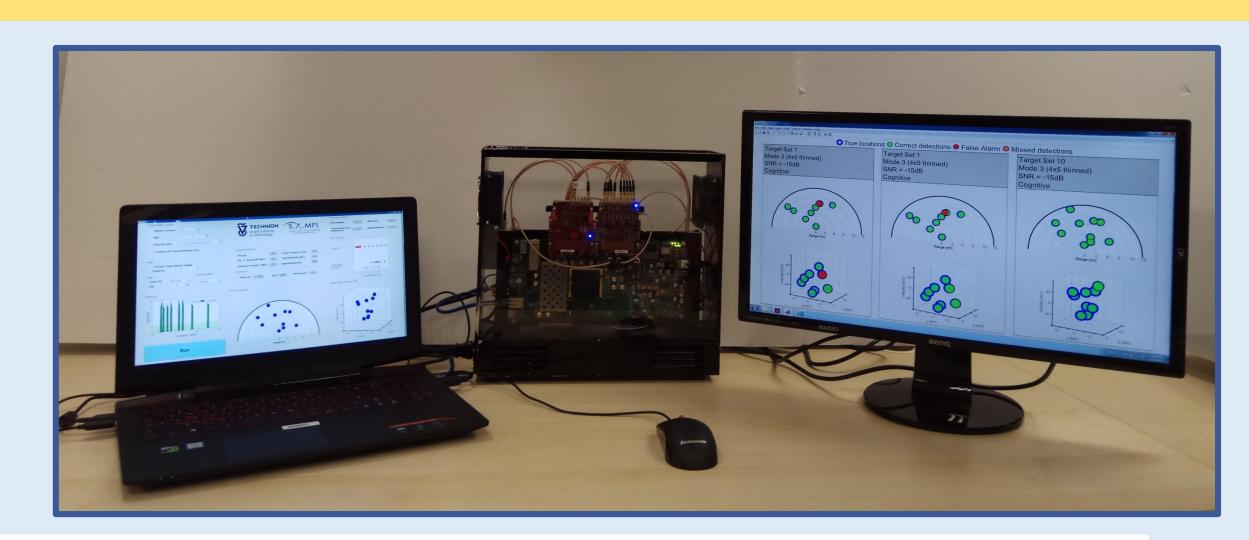
BPFs have ~30 dB stopband attenuation to mitigate subsampling noise

Digital Receiver

- Two 16-bit eight-channel digitizers for I and Q streams
- Sub-Nyquist sampling rate: 7.5 MHz/channel
- Signal BW with guardbands: 15 MHz/channel

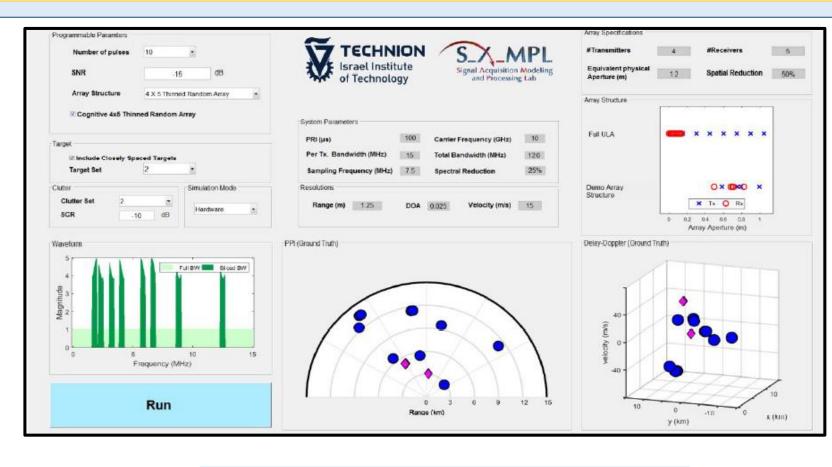


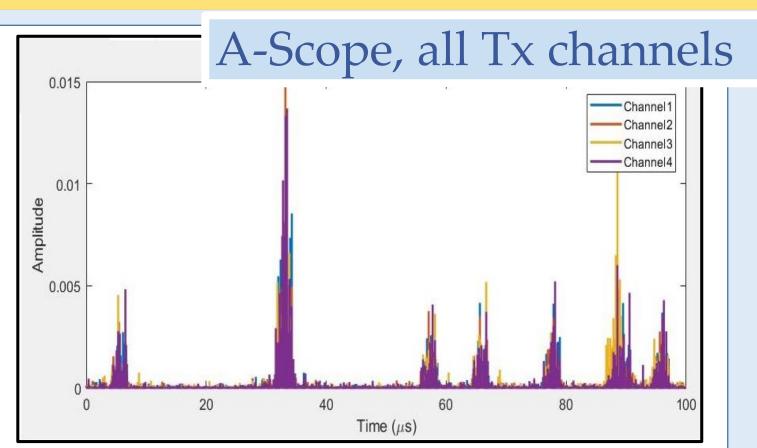
User Interface and Measurement Results (Mode 3, 4x5 Array)



Prototype with user control and measurement output

- Selectable scenarios, including closely-spaced targets
- Mode 3: 4x5 sub-Nyquist array resolution performance same as the virtual ULA (Mode 1)
- Mode 4: 8x10 sub-Nyquist array shows higher resolution performance than virtual 20x20 ULA





User Input Interface

