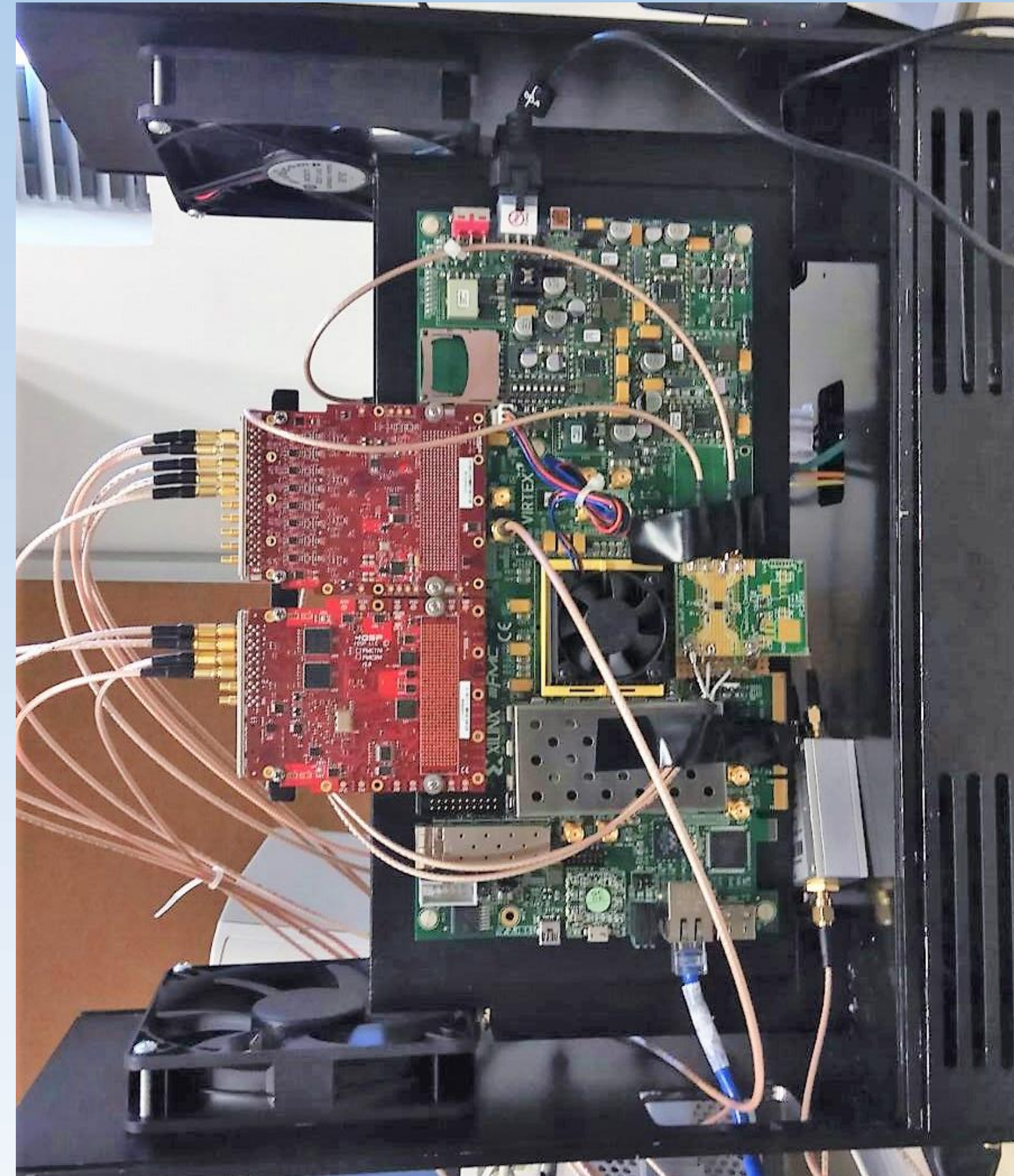


# Channel Estimation with Reduced RF Chains

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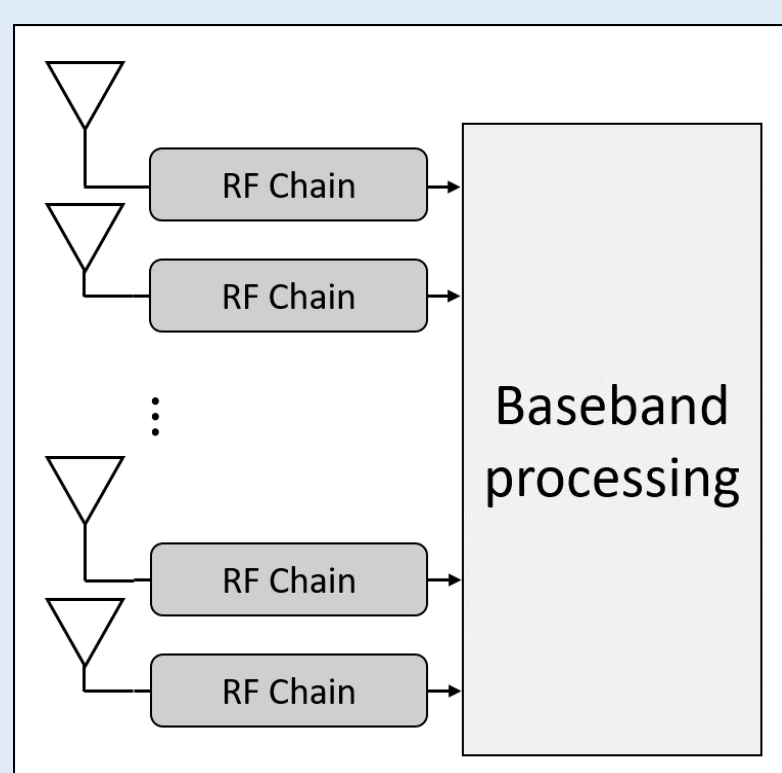
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## Theoretical Background

### Main Contributions

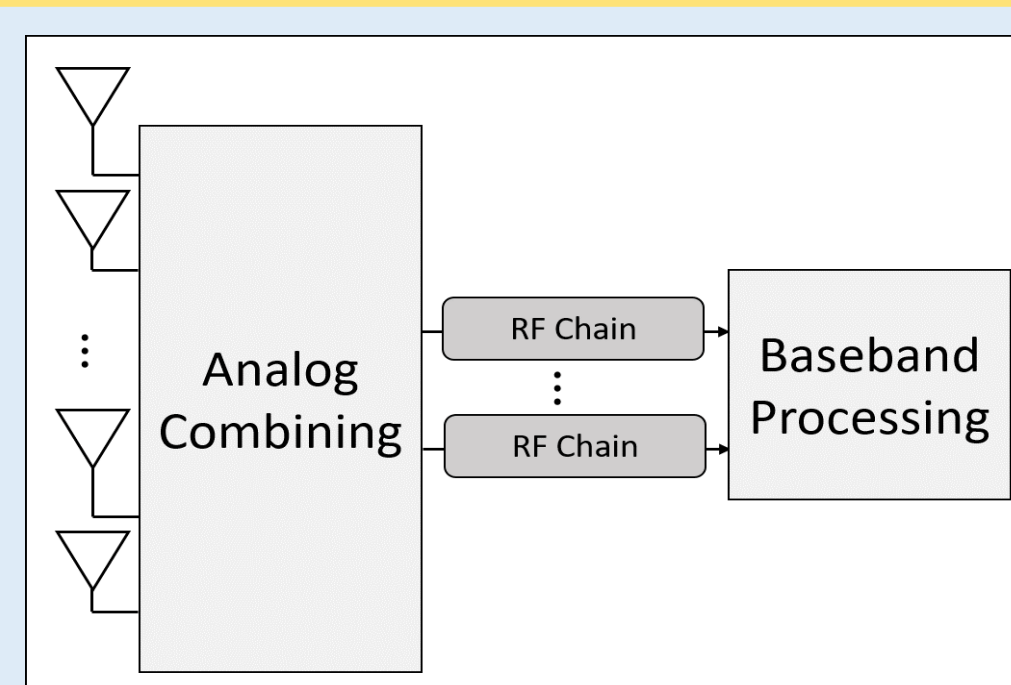
- This prototype demonstrates a **MIMO receiver with reduced number of RF chains**
- Employs 4 antennas and only 2 RF chains
- Demonstrates an analog combiner that consists of **controllable analog network** of phase shifters, gains and switches
- Performs MMSE channel estimation for Kronecker channel model in a multi-user MIMO scenario
- When the channel matrix is low rank the RF reduction does not increase the channel's MSE

### Conventional MIMO Receiver



- Each antennas is followed by a dedicated RF chain
- All the analog inputs from the antennas are accessible at the baseband
- All the signal processing operations are performed in baseband (fully-digital)

### Hybrid MIMO Receiver



- An analog combiner projects the high number of analog inputs from the antennas onto the low number of RF chains
- The analog combiner consists of a controllable network of phase shifters, gains and switches
- Only a low number of measurements are accessible in baseband

### Problem Formulation

- Multi-user MIMO scenario with 2 user terminals and a base station with 4 antennas and 2 RF chains
- The users transmit known, orthogonal pilot sequences of length  $\tau$  to the base station, over a TDD uplink channel
- Received signal at the baseband

- Kronecker channel model:

$$\mathbf{H} = \mathbf{R}_r^{\frac{1}{2}} \bar{\mathbf{H}} \mathbf{R}_t^{\frac{1}{2}}$$

- Receive side correlation
- White channel matrix with entries  $\sim \mathcal{N}(0,1)$
- Transmit side correlation
- Vectorized received signal:

$$\mathbf{y} = (\mathbf{S} \otimes \mathbf{W}) \mathbf{h}$$

- MMSE channel estimator

$$\hat{\mathbf{h}} = [\mathbf{R}_r \mathbf{S}^* (\mathbf{S} \mathbf{R}_r \mathbf{S}^*)^{-1} \otimes \mathbf{R}_t \mathbf{W}^* (\mathbf{W} \mathbf{R}_t \mathbf{W}^*)^{-1}]^{-1} \mathbf{y}$$

$$\mathbf{Y} = \mathbf{W} \mathbf{H} \mathbf{S}^T$$

Received digital signal

Analog combiner matrix

Channel matrix

Pilot sequences matrix

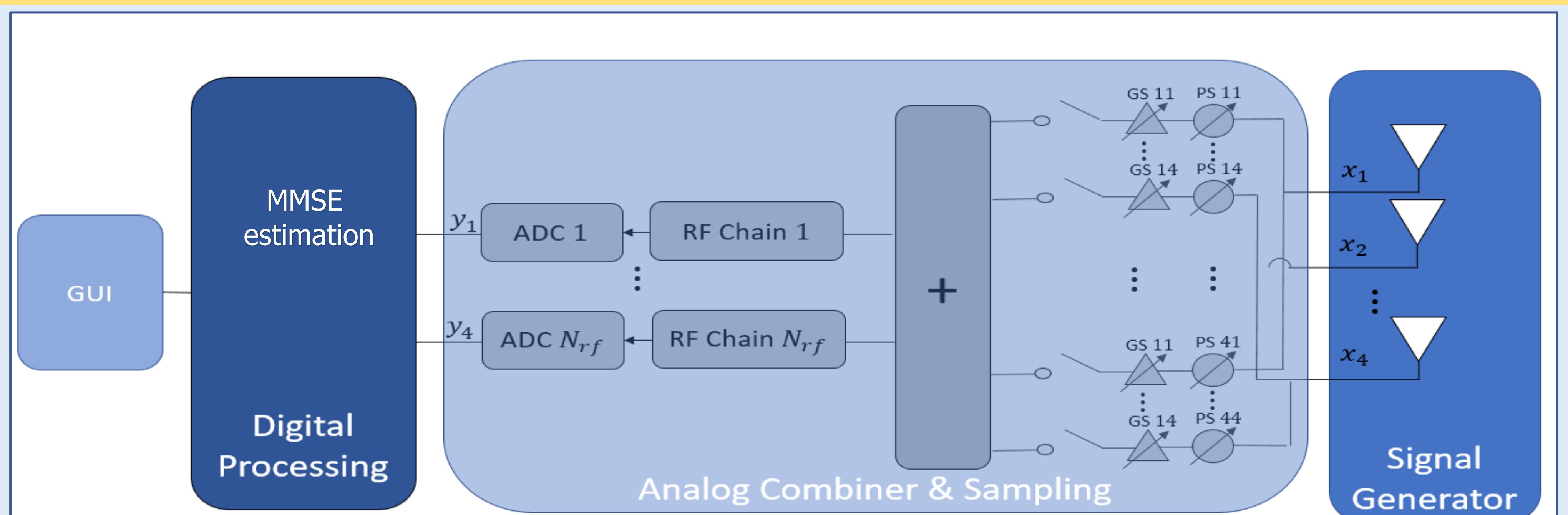
### Optimal Analog Combiner

- Due to the separable structure of the Kronecker model, an **optimal analog combiner** can be derived.
- [Stein and Eldar '2018]

$$\mathbf{W}_{opt} = \mathbf{T} \mathbf{U}_1^*$$

- $\mathbf{U}_1$  - First 2 eigenvectors of  $\mathbf{R}_r$
- $\mathbf{T}$  - Any  $2 \times 2$  invertible matrix

### Overview of Hardware Architecture



# Channel Estimation with Reduced RF Chains

## Technical Specification

Carrier Frequency -  $f_c = 1\text{GHz}$

Baseband BW = 125MHz – can be extended up to 2GHz

DAC - 4 output channels at 250MSPS

ADC – 4 input channels  
Sampling Rate – 250MHz for each channel

## Theoretical MSE

- For the suggested combiner  $W_{opt}$  the theoretic MSE in a noise free scenario is the sum of 2 smallest eigenvalues of the receive-side correlation:

$$\mathbb{E} [\|\hat{H} - H\|^2] = \lambda_3(\mathbf{R}_r) + \lambda_4(\mathbf{R}_r)$$

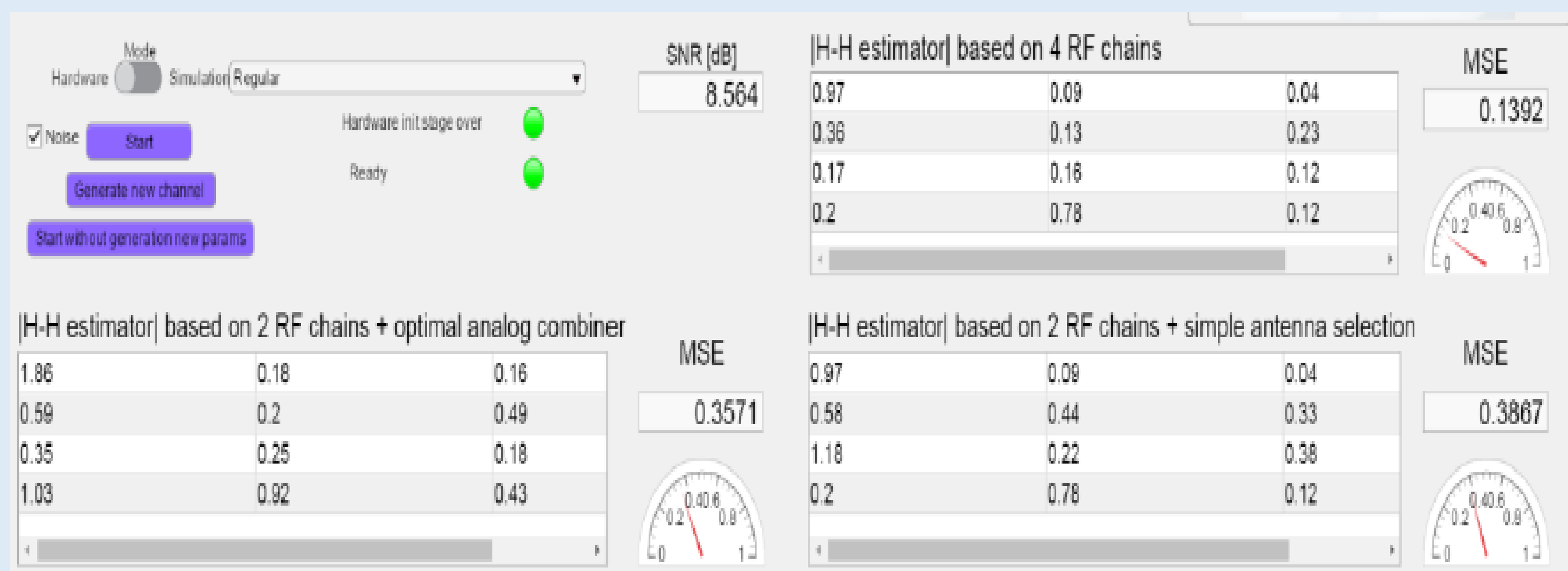
- Best case scenario:**  
 $\text{rank}(\mathbf{R}_r) \leq 2$ . In this case the MSE is:

$$\mathbb{E} [\|\hat{H} - H\|^2] = 0$$

- Worst case scenario:**  
 $\lambda_1(\mathbf{R}_r) = \lambda_2(\mathbf{R}_r) = \lambda_3(\mathbf{R}_r) = \lambda_4(\mathbf{R}_r) = a$ . In this case the optimal combining scheme is a simple “antenna-selection” scheme, and the MSE is:

$$\mathbb{E} [\|\hat{H} - H\|^2] = 2a$$

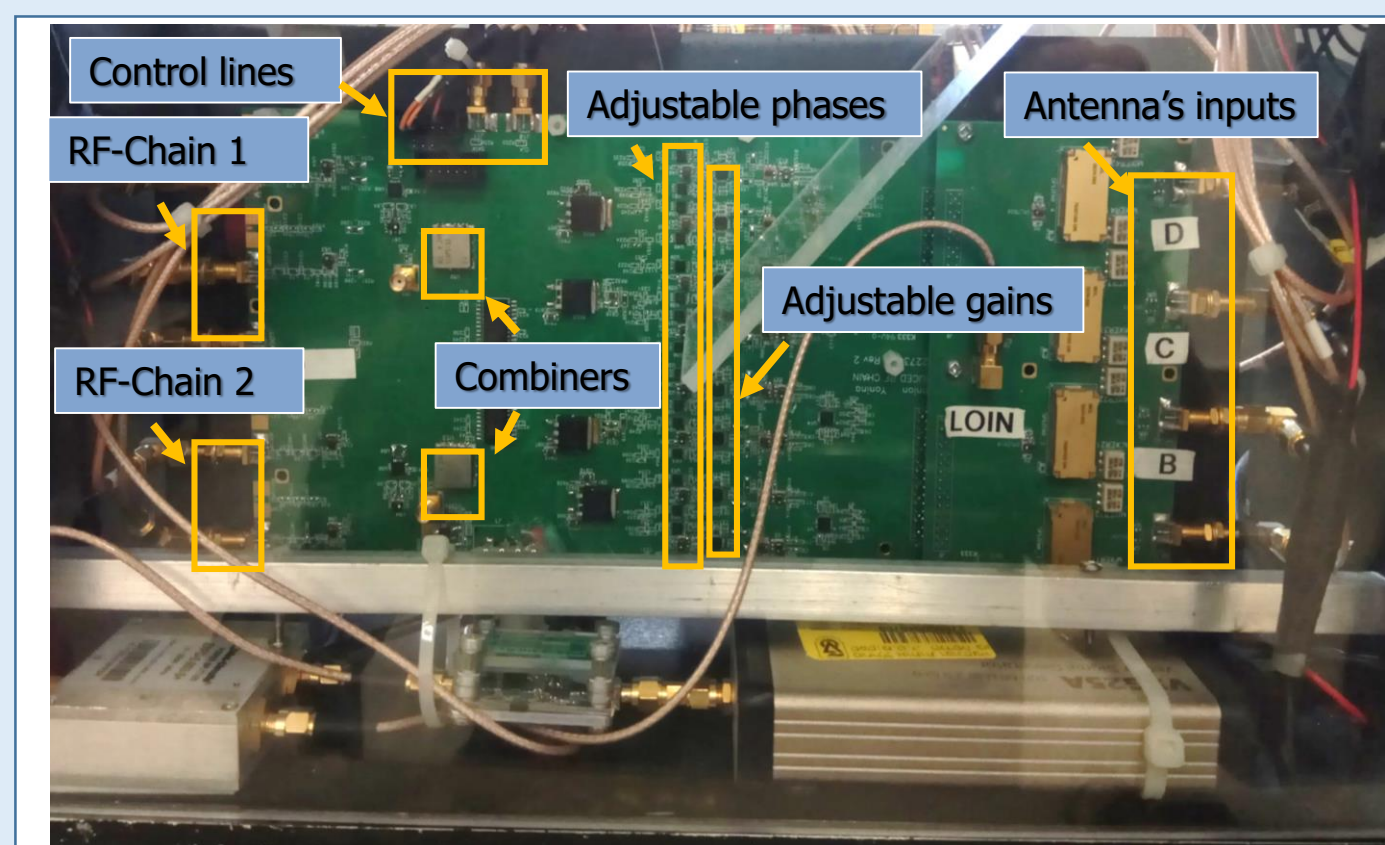
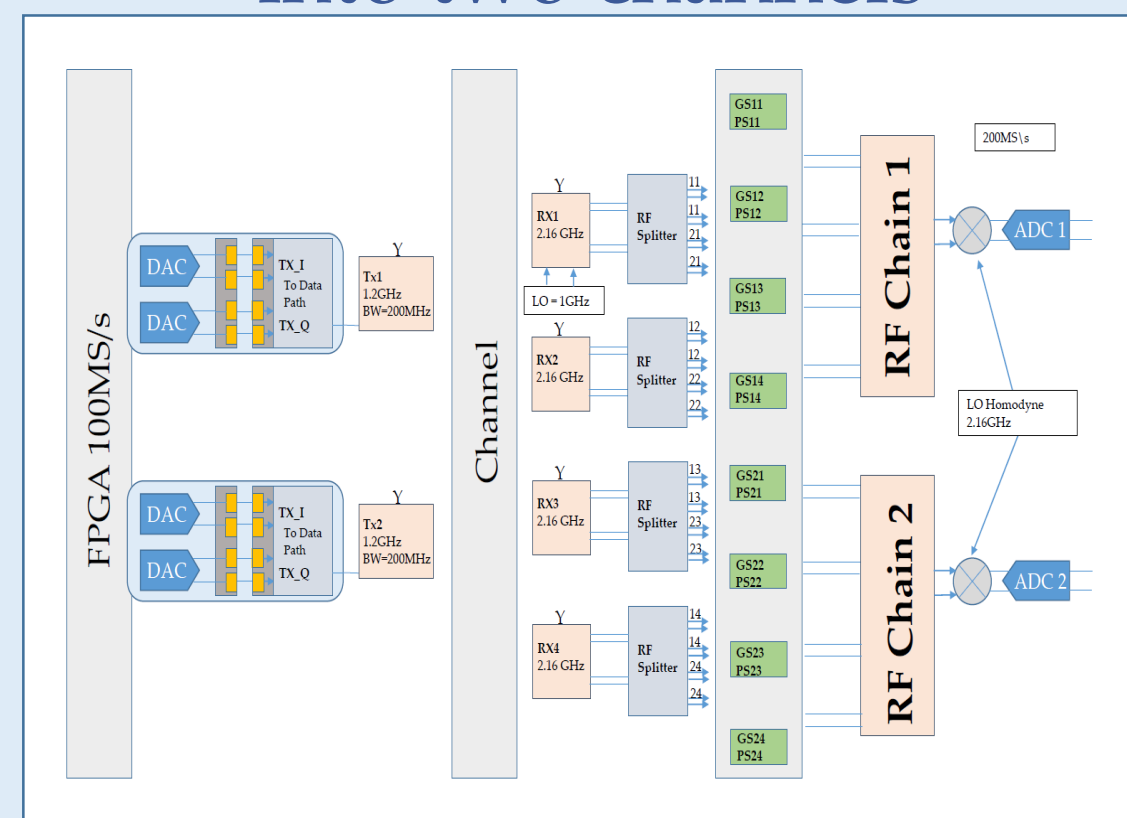
## Measurement Result



## Analog Pre-Processor (APP) Board

APP filters the receiver data coming from four channels into two channels

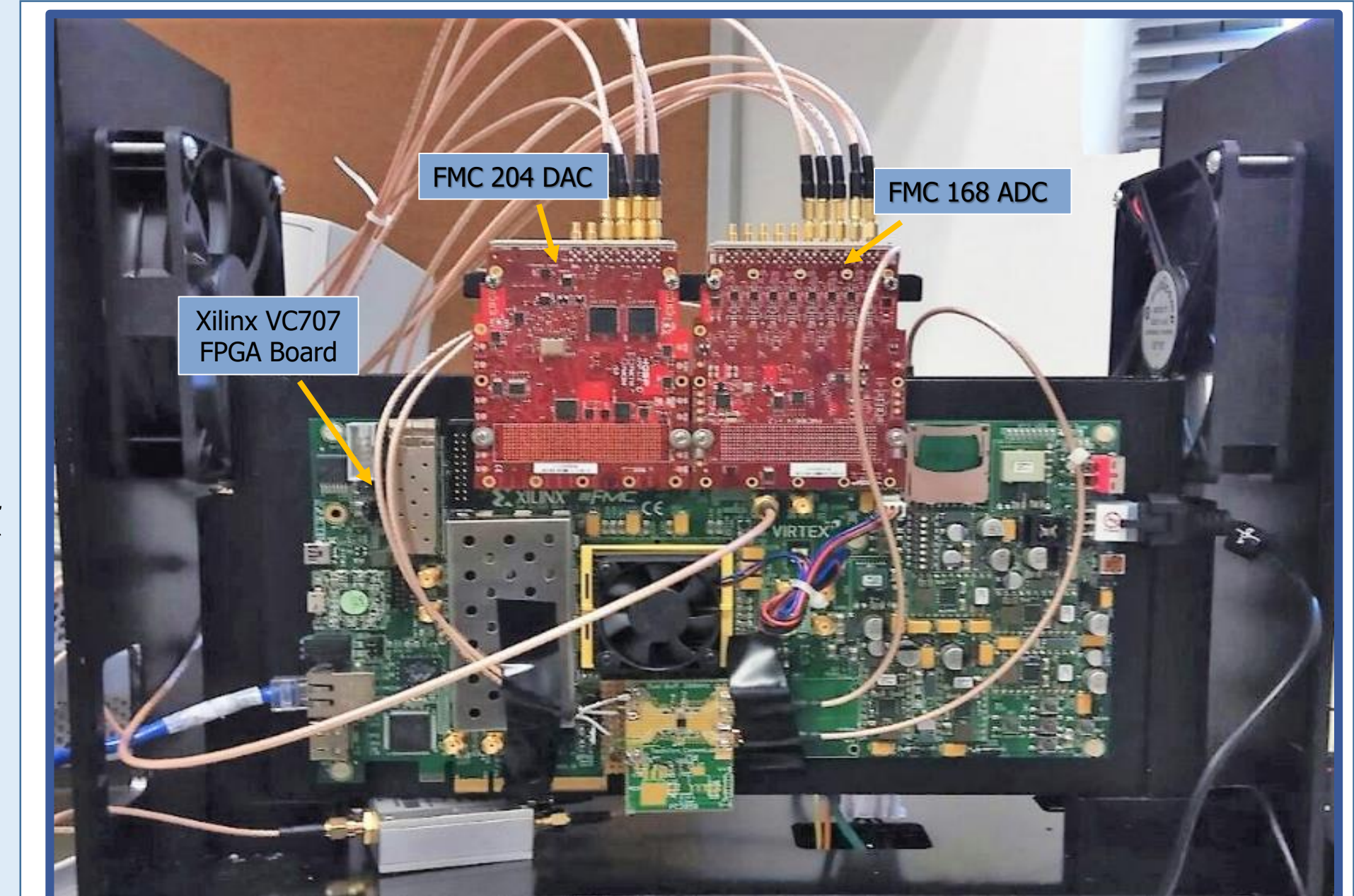
APP in a single chassis



Phase shifters and Amplitude shifters easily configured by FPGA controller

## Digital Receiver and Waveform Generator

- 16-bit 8-channel digitizer
- 16-bit 4-channel DAC for Waveform Generation
- Analog signal phase shift resolution -  $0.5^\circ$
- Amplitude modulation



## User Interface

