

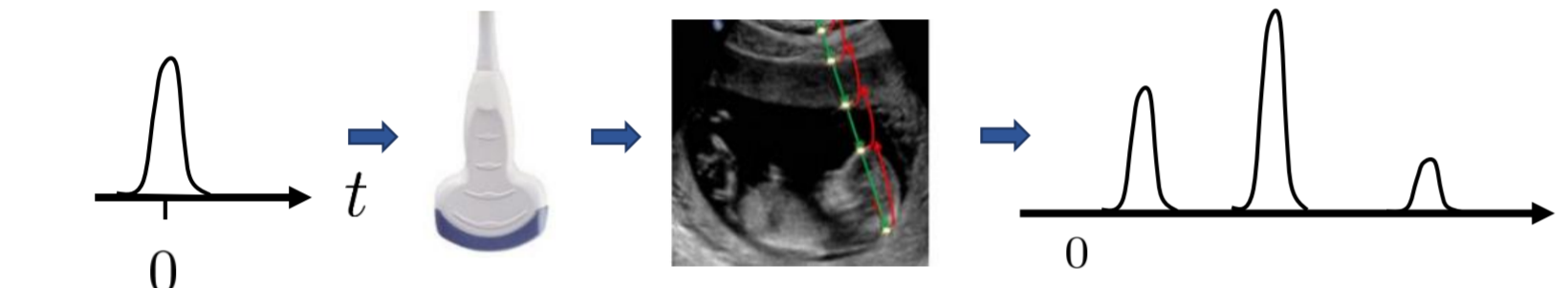
Efficient Coded Ultrasound Imaging Based on Frequency Domain Processing

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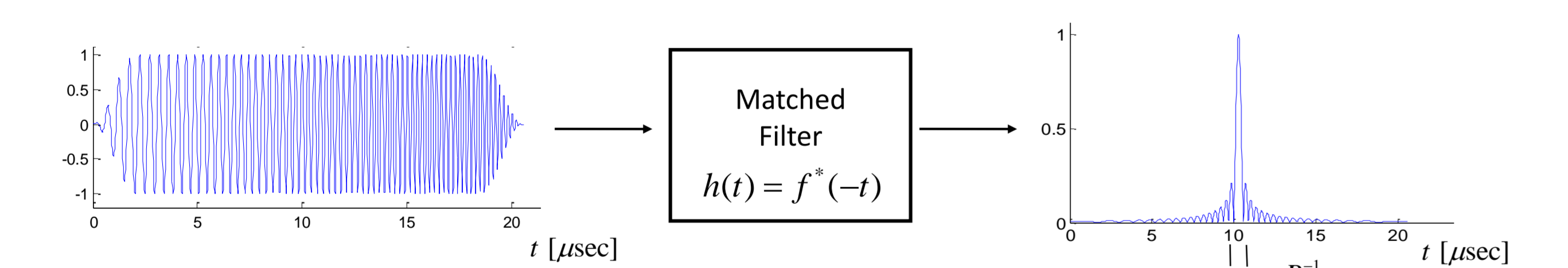
Motivation

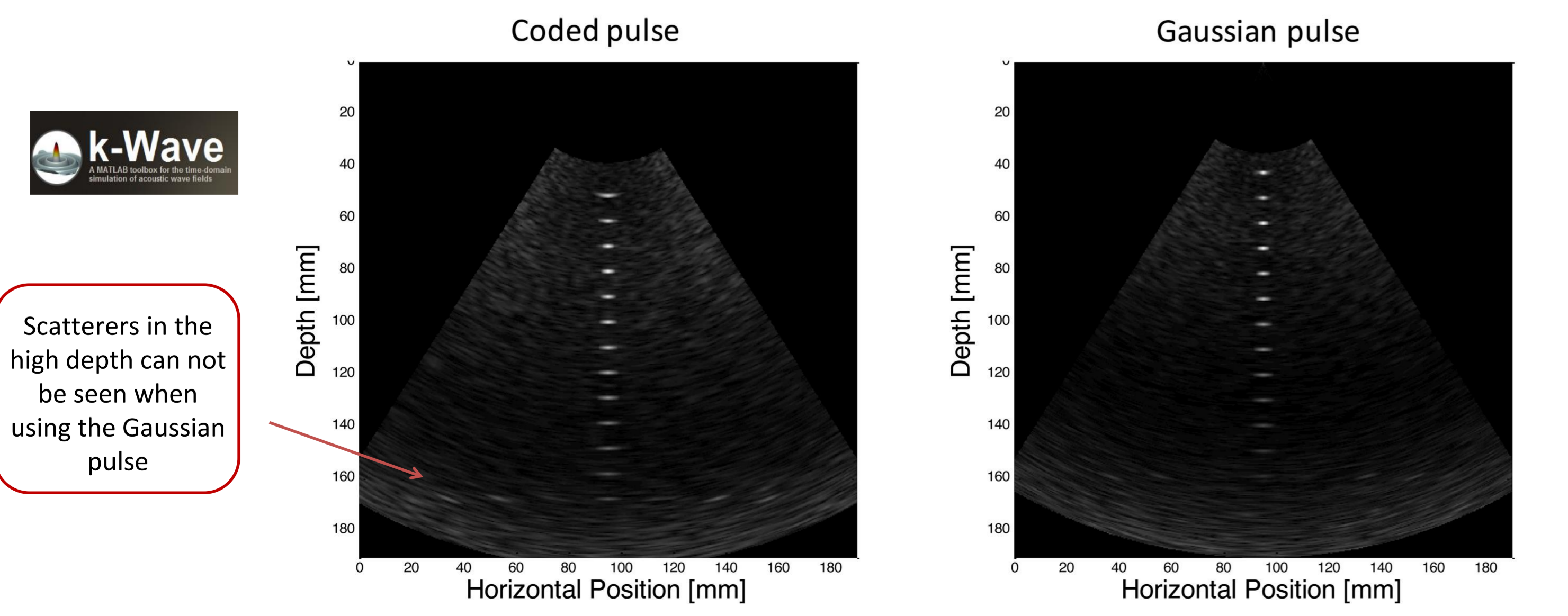
- Coded excitation (CE) and pulse compression increase the Tx energy and keep the same axial resolution
 - Improved SNR and imaging depth
 - Crucial for synthetic aperture and plane-wave based imaging modalities
- Bottleneck** – application to imaging with an array of elements due to increased computational complexity
- Our approach** is to introduce the pulse compression to computationally efficient frequency domain beamforming (FDBF)

Conventional Transmission

- In the standard approach transducer transmits a Gaussian pulse
 - Axial resolution is proportional to the pulse duration
 - SNR and imaging depth are defined by the transmitted energy
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- Transmitting more energy with the same pulse duration (resolution) requires higher intensity peak, which is limited by the FDA*

Coded Excitation Approach

- Coded signals are transmitted and compressed upon reception
 - An image line is comprised of pulse's autocorrelations
 - Linear frequency modulated (FM) signal:
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- The resolution is inversely proportional to B, the available system bandwidth determined by the transducer properties
- CE breaks the link between axial resolution and the transmitted energy**



Beamforming (BF) and Pulse Compression

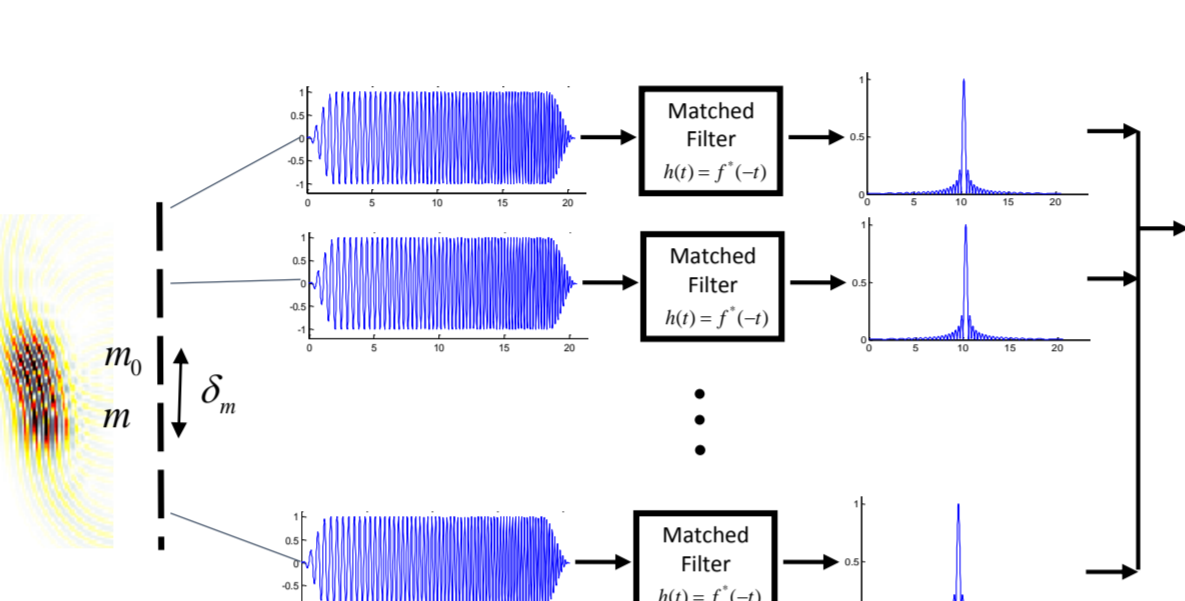
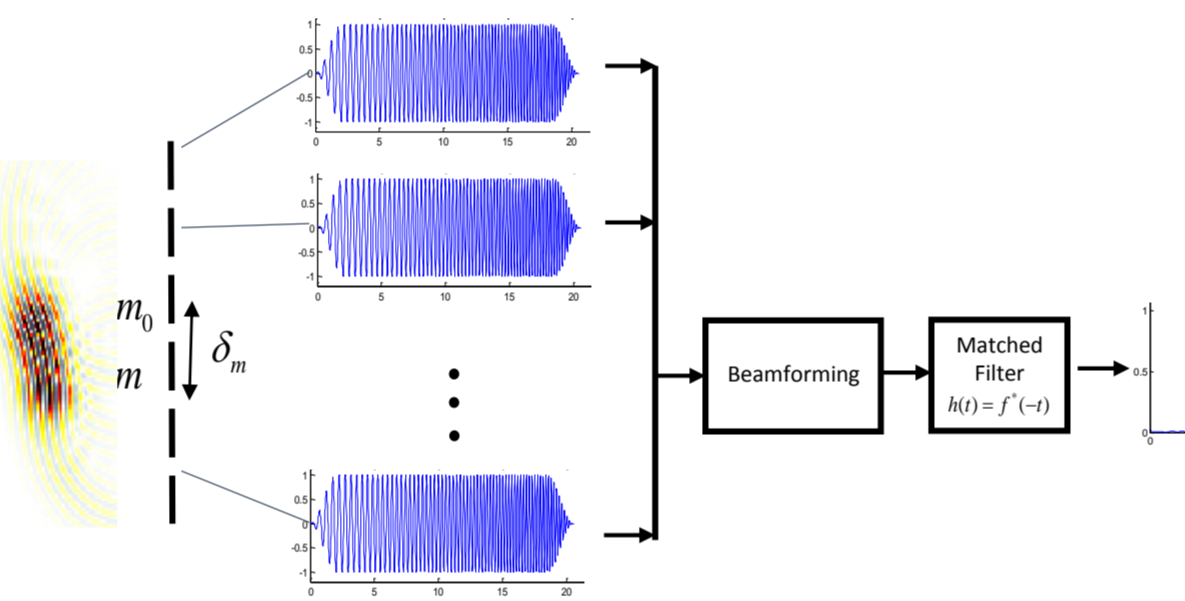
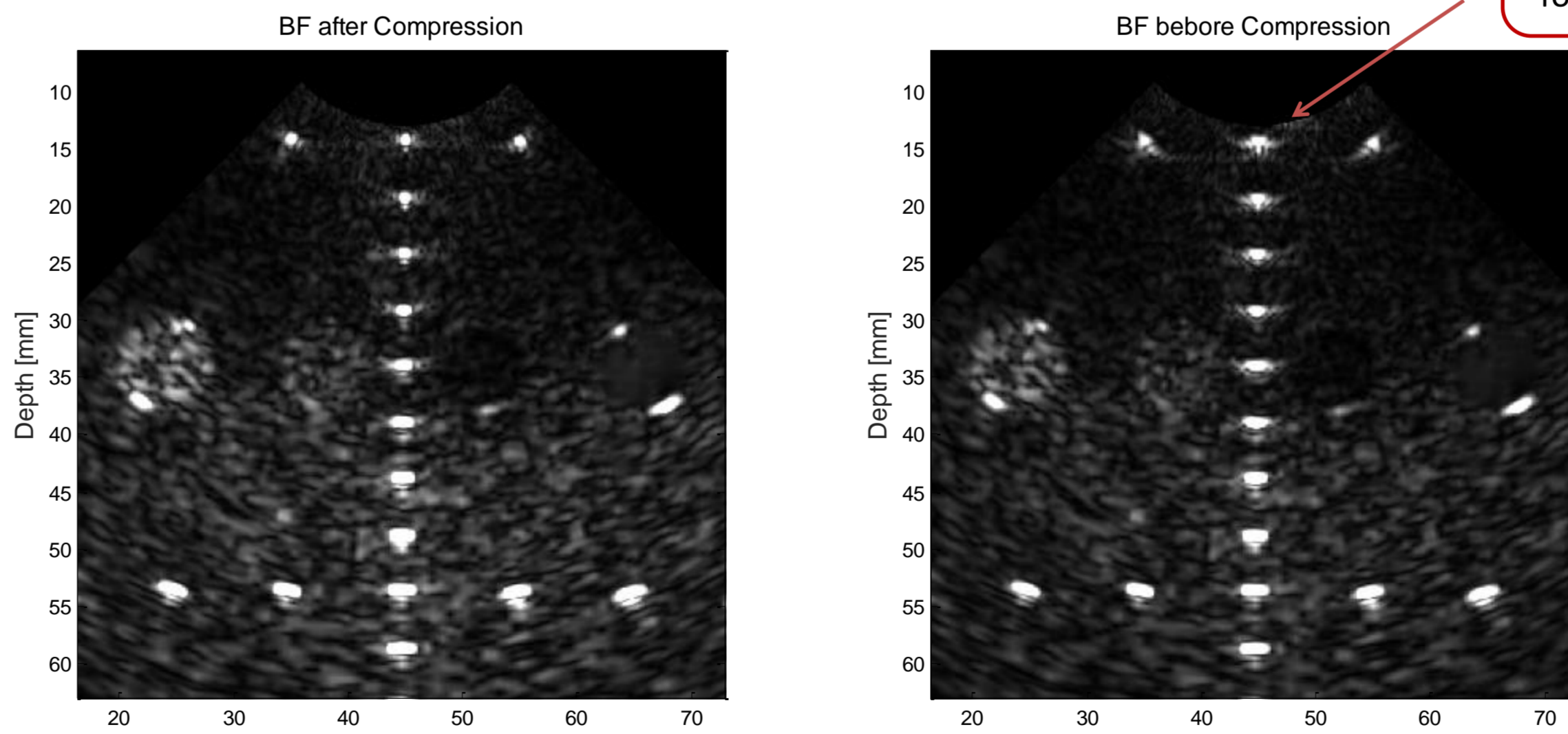
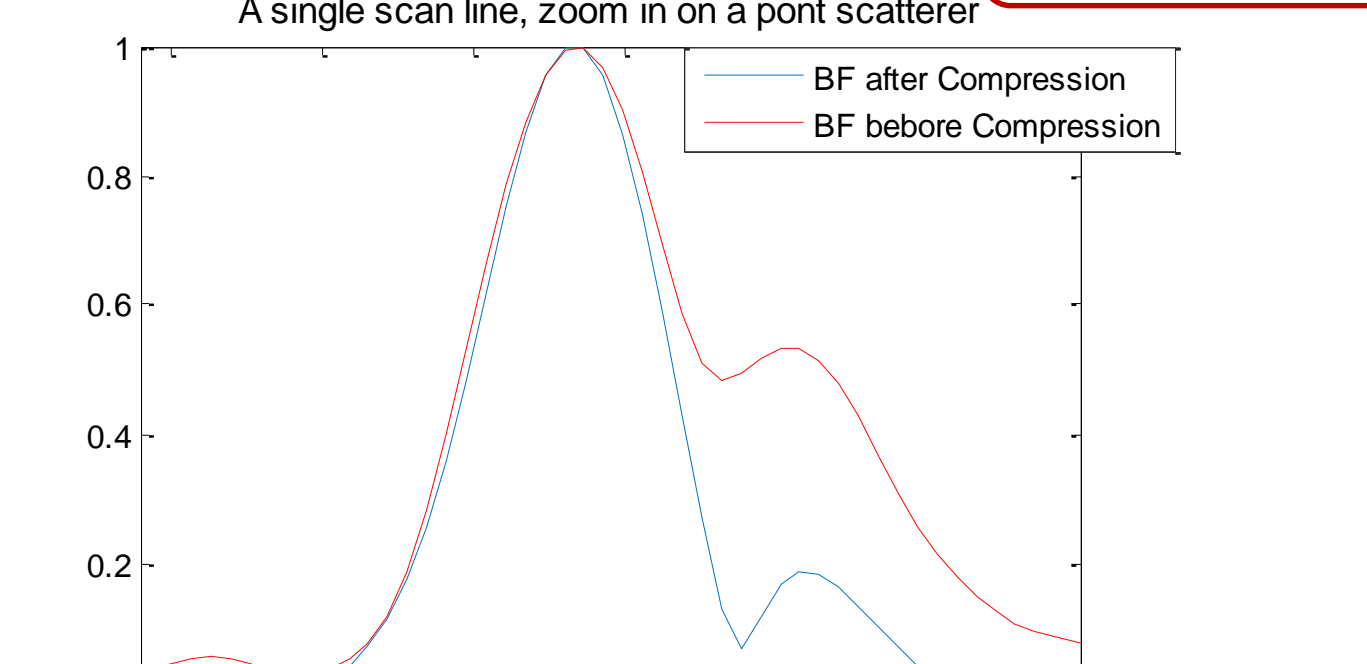
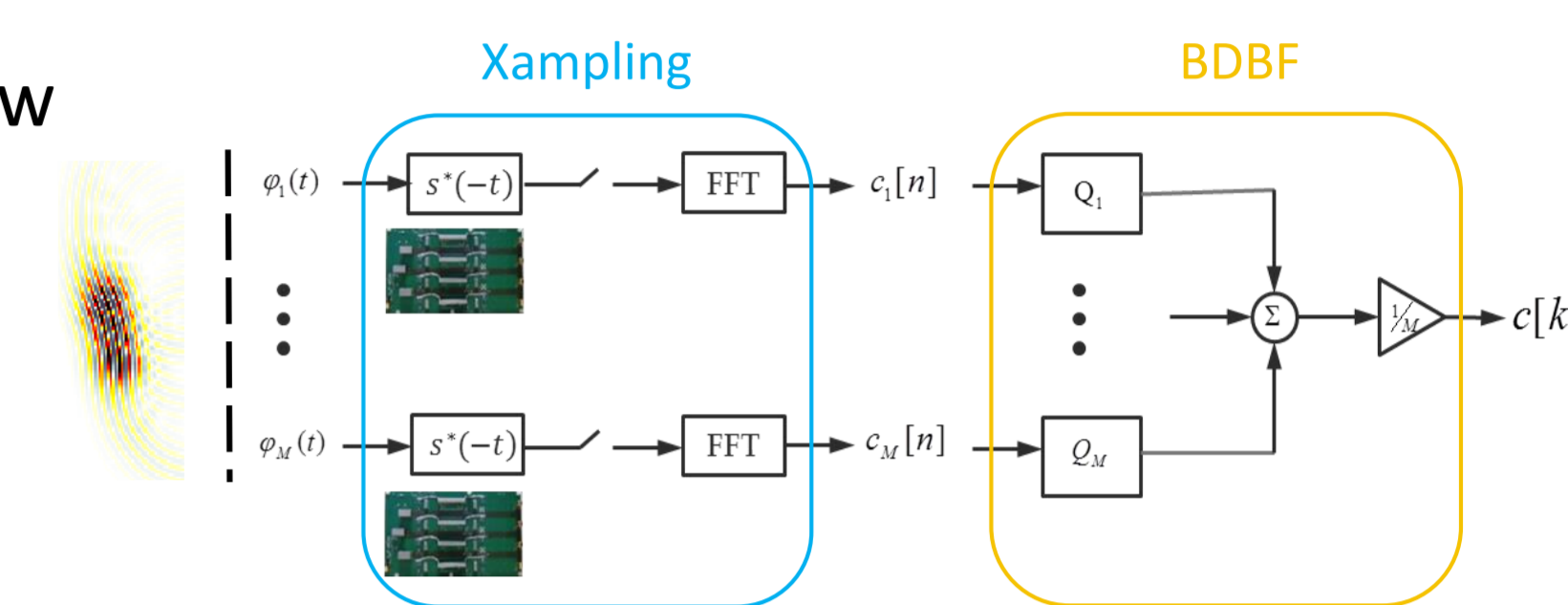
- Optimal processing requires compressing every channel prior to BF
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- $$\Phi_{CE} = \frac{1}{M} \sum_{m=1}^M \{\varphi_m * h\}(\tau_m(t; \theta))$$
- $$\tau_m(t; \theta) = \frac{1}{2} \left(t + \sqrt{t^2 - 4(\delta_m/c)t \sin \theta + 4(\delta_m/c)^2} \right)$$
- The computational complexity is vastly increased by filtering each detected signal **restricting the use of CE in array imaging**
- An obvious solution – beamforming before pulse compression
 - Saving $(N_e - 1) \left(\frac{3}{2} (N_s + N_h) \log(N_s + N_h) + N_s + N_h \right)$
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- Non-linear time dependent delays distort the phases of the coded signals
- $$\Phi_{CE_{pre}} = \left\{ \frac{1}{M} \sum_{m=1}^M \varphi_m \left(\tau_m(t; \theta) \right) \right\} * h(t)$$

Image Degradation due to Compression Error

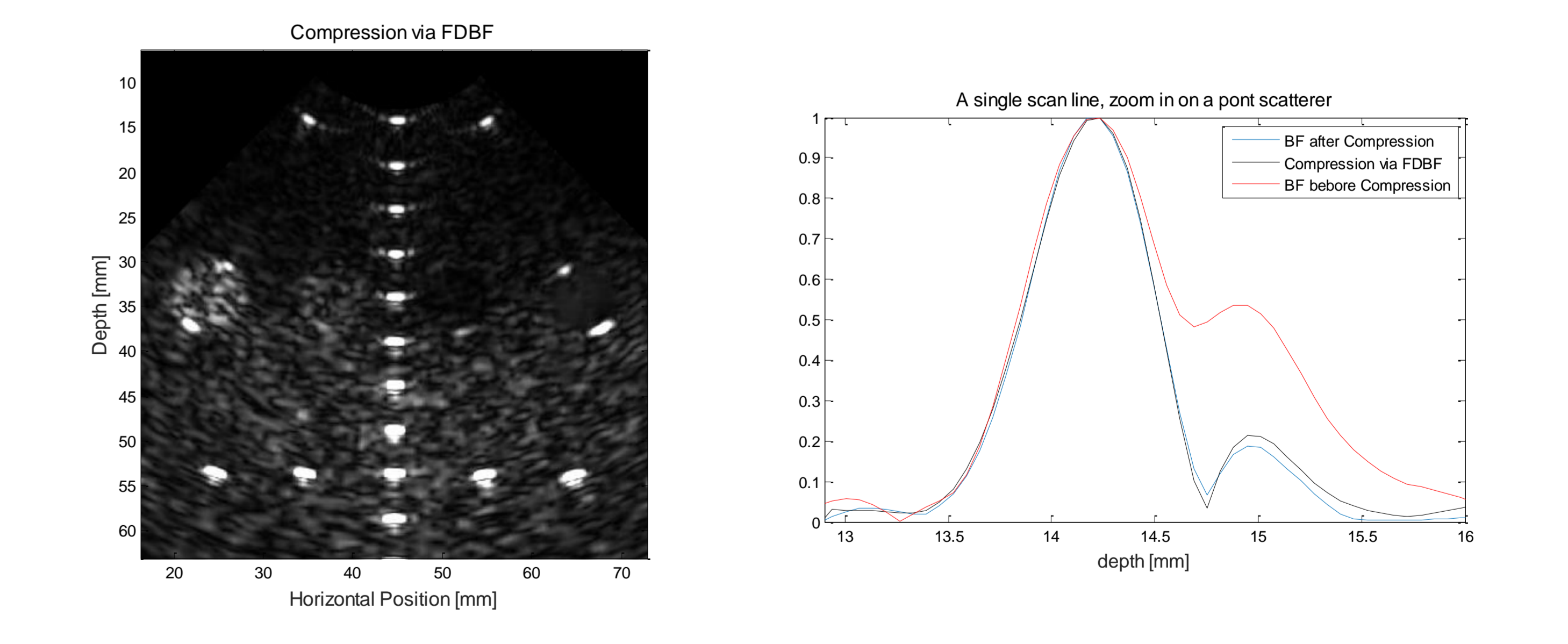
- Non-linear phase distortion affects the properties of the compressed output and degrades the image quality
- $$\Phi_{CE} = \frac{1}{M} \sum_{m=1}^M \{\varphi_m * h\}(\tau_m(t; \theta)) \neq \Phi_{CE_{pre}} \left\{ \frac{1}{M} \sum_{m=1}^M \varphi_m(\tau_m(t; \theta)) \right\} * h(t)$$
- The errors are more prominent for low depth
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- Main lobe ~13% wider
Sidelobe ~9 dB higher

- Other existing solutions
- Limiting the coded signal duration ($T < 64/f_0$)
Drawback – limited SNR improvement
 - Imaging several zones with different codes
Drawback – reduced frame rate
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Frequency Domain Beamforming and Pulse Compression

- Each channel is sampled at a low rate using Xampling approach
 - Beamforming is performed directly in frequency
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- Standard Imaging
- $$\Phi(t; \theta) = \frac{1}{M} \sum_{m=1}^M \varphi_m(\tau_m(t; \theta))$$
- $$c[k] = \frac{1}{M} \sum_{m=1}^M \sum_{n=-N_1}^{N_2} c_m[k-n] Q_{k,m;\theta}[n]$$
- Coded Imaging
- $$\Phi_{CE}(t; \theta) = \frac{1}{M} \sum_{m=1}^M \{\varphi_m * h\}(\tau_m(t; \theta))$$
- $$c_{CE}[k] = \frac{1}{M} \sum_{m=1}^M \sum_{n=-N_1}^{N_2} c_m[k-n] h[k-n] Q_{k,m;\theta}[n]$$
- $$= \frac{1}{M} \sum_{m=1}^M \sum_{n=-N_1}^{N_2} c_m[k-n] Q_{k,m;\theta}^{CE}[n]$$
- $$\varphi_m^{CE}(t) = \{\varphi_m * h\}(t) \rightarrow c_m^{CE}[k] = c_m[k]h[k]$$
- The set $\{Q_{k,m;\theta}^{CE}[n]\}$ includes the Fourier series coefficients of the matched filter
 - Performs beamforming and pulse compression simultaneously in the frequency domain
- Pulse compression is performed in frequency together with beamforming and does not require any additional computational load*

Experimental Results



- Verasonics Vantage 256
64 element probe, $f_0 = 2.98\text{MHz}$
 $f_c = 11.9\text{MHz}$
- Computational complexity**
- Time domain processing $N_{time} = MN_s + M \left(\frac{3(N_s + N_h)}{2} \log(N_s + N_h) + N_s + N_h \right)$
 - Frequency domain processing $N_{frequency} = MN_Q + \frac{N_s}{2} \log N_s$
- 16 fold reduction in the number of multiplications**