



$$c = \int_a^b \log_2 \left( 1 + \frac{S(f)}{N(f)} \right) df$$

$$\hat{x} = \arg \min_{x \in \mathbb{R}^N} \|x\|_1 + \lambda \|y - Ax\|_2^2$$



# A Software Prototype of Compressed Ultrasound imaging over WIFI

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## Motivation and Contributions

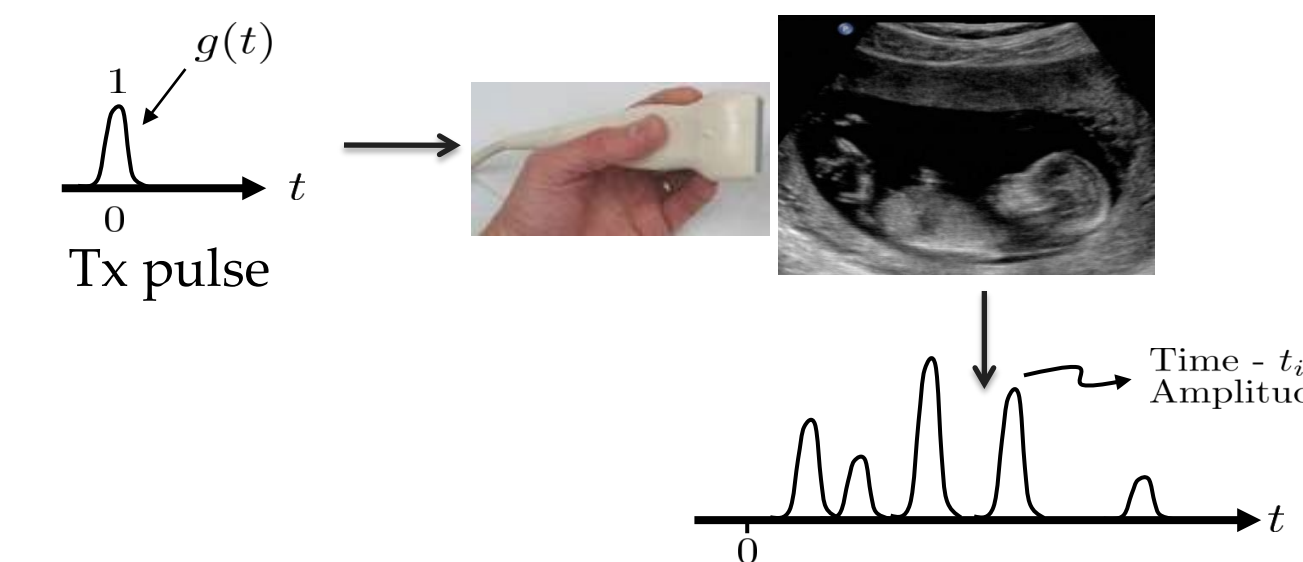
- ❑ Ultrasound imaging is one of the most common medical imaging technique
- ❑ The commercially used US machines are relatively large, with limited image resolution and contrast
- ❑ Uncomfortable cable management
- ❑ Reducing sampling rate and data size while ensuring high image quality can make US imaging the chosen imaging method in many medical scenarios
- ❑ We present a low rate US imaging technique with sub sampled channel data transferred over WIFI

## Ultrasound – delay and sum beamforming

- ❑ Echoes result from scattering in the tissue
- ❑ The image is formed by identifying the scatterers
- ❑ To increase SNR and resolution an antenna array is used
- ❑ SNR and resolution are improved through beamforming by introducing appropriate time shifts to the received signals

$$\Phi(t; \theta) = \frac{1}{M} \sum_{m=1}^M \varphi_m \left( t - \frac{1}{2} \left( t - \sqrt{t^2 - 4(\delta_m/c) \sin \theta + 4(\delta_m/c)^2} \right) \right)$$

- ❑ Steers the beam in a desired direction and focuses it in the ROI
- ❑ **Requires high sampling rates and large data processing rates**



## Compressed beamforming

- ❑ Each individual trace is buried in noise and has no structure
- ❑ Structure exists only after beamforming which improves resolution/SNR
- ❑ How can we perform beamforming on low rate data?

**Enables beamforming from low rate samples**  
**Key idea: Perform beamforming in frequency**

Logic:

1. BMF signal is a stream of pulses => can be recovered from a small number of  $c_k$
2. Small number of  $c_k$  requires only a small number of  $\varphi_m[n]$

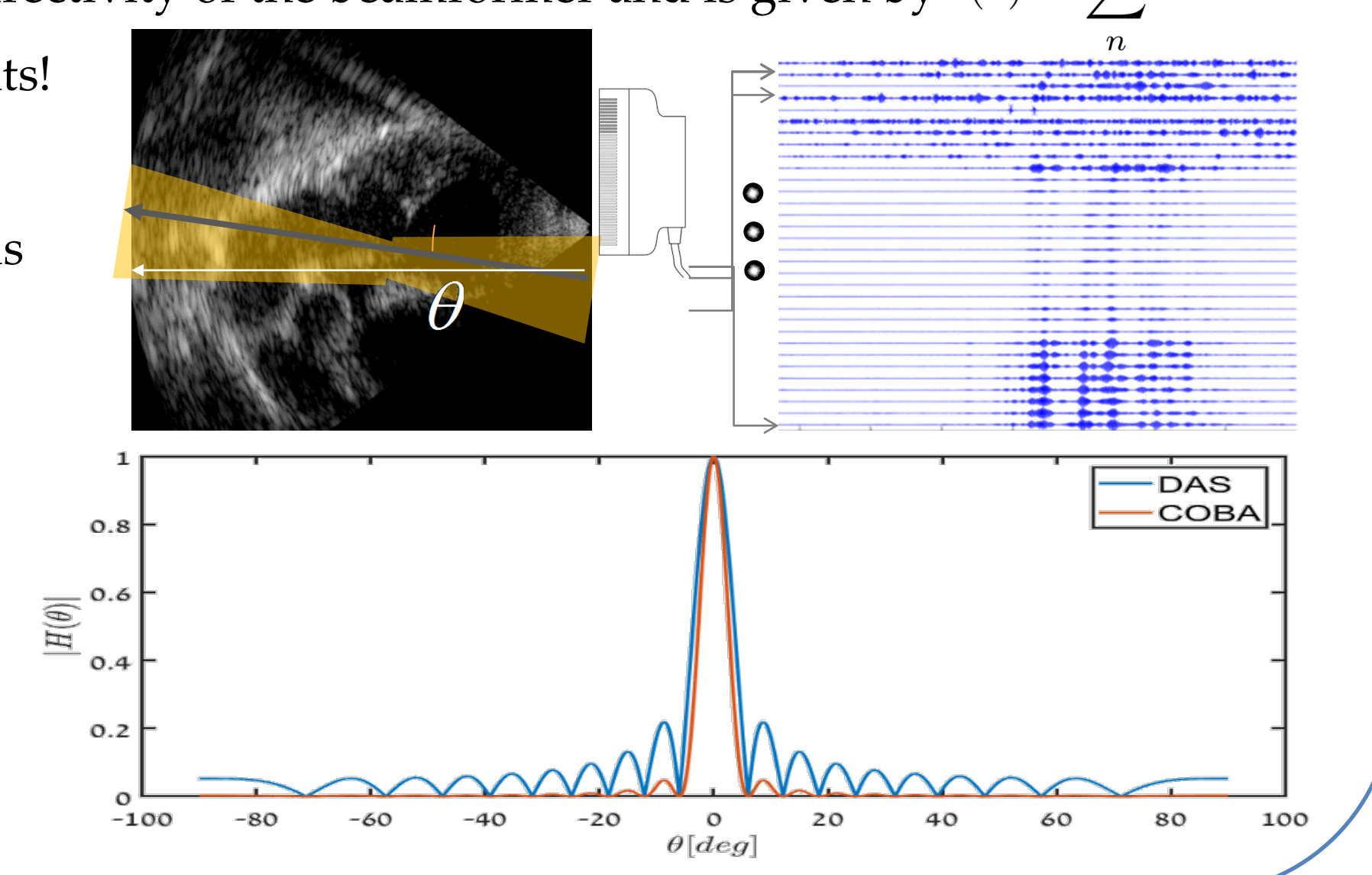
$$c_k = \frac{1}{M} \sum_{m=1}^M \sum_n \varphi_m[n] Q_{k,m,\theta}[k-n]$$

Fourier coefficient of BMF signal  $Q_{k,m,\theta}[n]$  are the Fourier coefficients of  $q_{k,m,\theta}(t; \theta)$

$$q_{k,m,\theta}(t; \theta) = I_{(t, c_m/c, \tau, \theta)}(t) \left( 1 + \frac{(\delta_m/c)^2 \cos^2 \theta}{(t - \delta_m/c) \sin \theta} \right) \times \exp \left\{ i \frac{2\pi}{T} k \frac{\delta_m/c - t \sin \theta}{(t - \delta_m/c) \sin \theta} \right\}$$

## Convolutional Beamforming – Array Selction

- ❑ Image quality is determined by the beampattern  $H(\theta)$  which represents the directivity of the beamformer and is given by  $H(\theta) = \sum_n e^{-2\pi j \frac{\sin(\theta)}{\lambda} nd}$
- ❑ Convolutional beamforming - same and better beampattern with less elements!
- ❑ Output of a standard delay and sum beamformer:  
 $\Phi(t; \theta) = \sum_{n=-N}^N \hat{\phi}_n(t; \theta)$ , where  $M = 2N + 1$ ,  $\hat{\phi}_n$  are the delayed signals
- ❑ Convolutional beamformer (COBA):



- ❑ Compute  $y_n = \sqrt{(\hat{\phi}_n)} \text{sign}(\hat{\phi}_n)$
- ❑ Convolve  $\mathbf{s} = \mathbf{y} * \mathbf{y}$
- ❑ Sum  $b = \sum_{n=-2N}^{2N} s_n$
- ❑ The resulting beampattern is:  $H(\theta) = \sum_{n=-2N}^{2N} (2N - |n|) e^{-2\pi j \frac{\sin(\theta)}{\lambda} nd}$

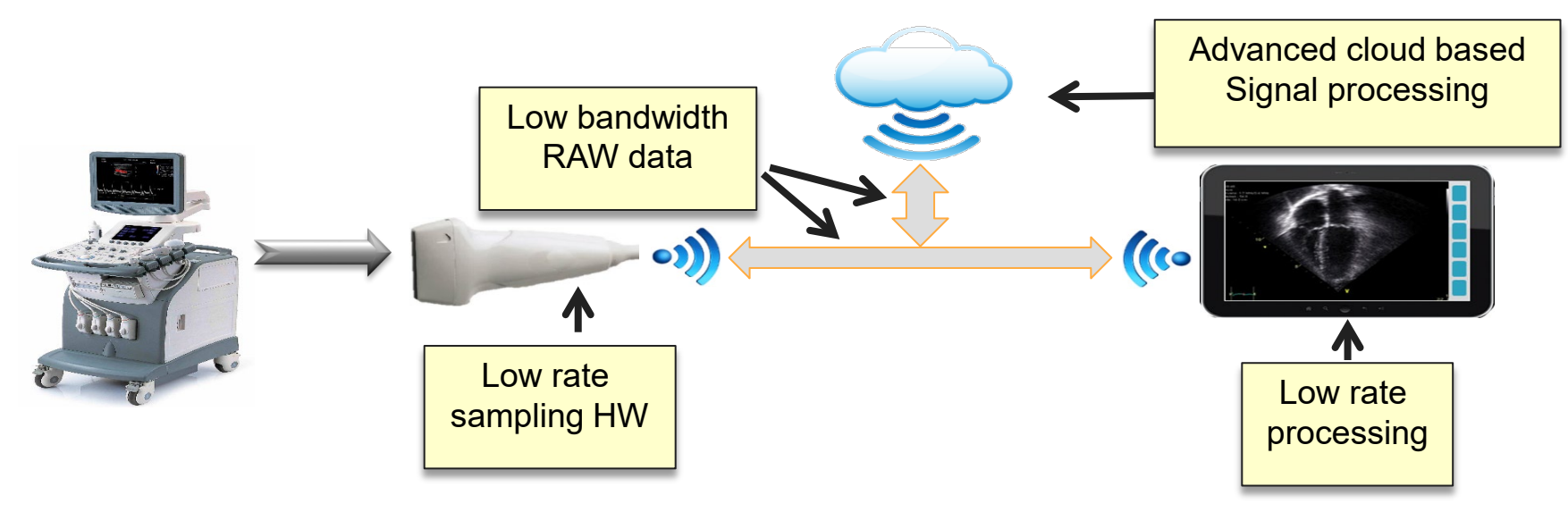
## Compressed Frequency COBA

- ❑ Reconstruct the sparse convolutionally beamformed signal from only a portion of the bandwidth of the signal received by sparse set of array elements
- ❑ Acquire non delayed data,  $\varphi_m(t; \theta)$ , by sparse set of elements
- ❑ Compute:  $y_m = \sqrt{(\hat{\phi}_m)} \text{sign}(\hat{\phi}_m)$
- ❑ Compute Fourier transform,  $\mathcal{F}\{y_m(t; \theta)\}$ , and apply delay in frequency domain:  $\hat{c}_m[k] = \frac{1}{M} \sum_n c_m[k-n] Q_{k,m,\theta}[n]$
- ❑ Calculate the Fourier coefficients of the convolutionally beamformed signal using the convolution theorem and 2-D IFFT:  $\hat{c}[k]_{COBA} = N \sum_{l \in S_{\hat{c}}} (\hat{c}_l * \hat{c})[k]$
- ❑ Reconstruct the signal based on the square of the square of the transmitted pulse

$$\min_b \|b\|_1 \quad s.t. \quad \|GDb - \hat{c}\|_2 \leq \epsilon \quad \text{NESTA} \rightarrow \hat{\Phi}_{COBA}(t) = \sum_{s=1}^S \left( \sum_{l \in \hat{U}} \sum_{m \in \hat{U}} a_{s,l} a_{s,m} \right) h^2(t - t_s) = \sum_{s=1}^S b_s g(t - t_s)$$

## From Vision to Implementation

- ❑ Sample within the probe
- ❑ Transmit channel data
- ❑ Recover in any imaging platform – tablet, cellular ...
- ❑ Send to cloud for advanced signal processing and AI methods



## Software + WIFI

